

# **BEE402-ELECTRICAL NETWORK ANALYSIS AND SYNTHESIS**

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# Property 1. L-C immittance function

- ▶ 1.  $Z_{LC}(s)$  or  $Y_{LC}(s)$  is the ratio of odd to even or even to odd polynomials.

- ▶ Consider the impedance  $Z(s)$  of passive one-port network.

$$Z(s) = \frac{M_1(s) + N_1(s)}{M_2(s) + N_2(s)} \quad (\text{M is even N is odd})$$

As we know, when the input current is  $I$ , the average power dissipated by one-port network is zero:

$$\text{Average Power} = \frac{1}{2} \text{Re}[Z(j\omega)] |I|^2 = 0$$

$$\text{EvZ}(s) = \frac{M_1(s)M_2(s) - N_1(s)N_2(s)}{M_2(s)^2 + N_2(s)^2} = 0$$

$$M_1(j\omega)M_2(j\omega) - N_1(j\omega)N_2(j\omega) = 0$$

$$M_1 = 0 = N_2 \quad \text{OR} \quad M_2 = 0 = N_1$$

$$Z(s) = \frac{M_1}{N_2}, Z(s) = \frac{N_1}{M_2}$$

**Z(s) or Y(s) is the ratio of even to odd or odd to even!!**

# Property 2. L-C immittance function

- ▶ 2. The poles and zeros are simple and lie on the  $j\omega$  axis.

$$Z(s) = \frac{M_1}{N_2}, Z(s) = \frac{N_1}{M_2}$$

- ▶ Since both M and N are Hurwitz, they have only imaginary roots, and it follows that the poles and zeros of Z(s) or Y(s) are on the imaginary axis.

- ▶ Consider the example

$$Z(s) = \frac{a_4 s^4 + a_2 s^2 + a_0}{b_5 s^5 + b_3 s^3 + b_1 s}$$

$$Z(s) = \frac{a_4s^4 + a_2s^2 + a_0}{b_5s^5 + b_3s^3 + b_1s}$$

In order for the impedance to be positive real  $\rightarrow$  the coefficients must be real and positive.  $j\omega$

Impedance function **cannot have multiple poles or zeros** on the axis.

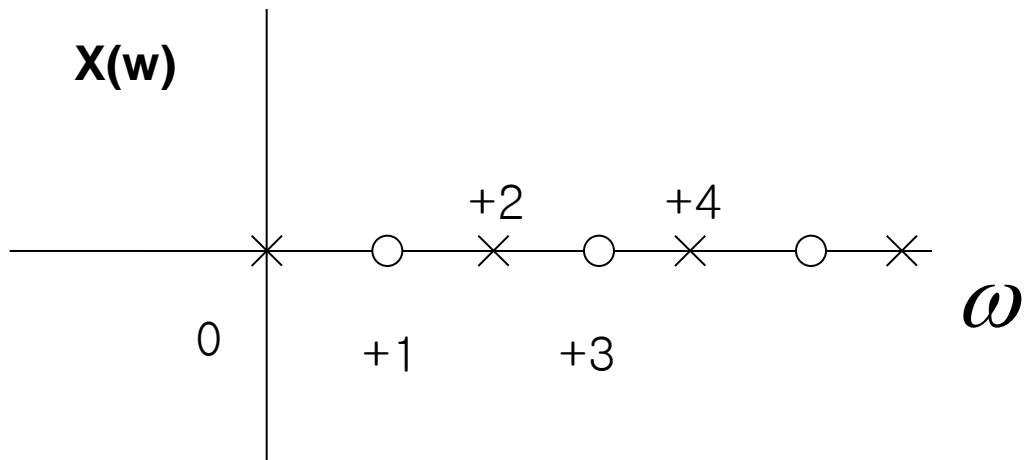
The highest powers of the numerator and the denominator polynomials can differ by, at most, unity.

$\infty$

Ex) highest order of the numerator :  $2n \rightarrow$  highest order of the denominator can either be  $2n-1$  (simple pole at  $s = \infty$ ) or the order can be  $2n+1$  (simple zero at  $s = \infty$ ).

# Property 3. L-C immittance function

- ▶ 3. The poles and zeros interlace on the  $j\omega$  axis.



Highest power:  $2n$   $\rightarrow$  next highest power must be  $2n-2$

They cannot be missing term. Unless?

$$\text{if } b_5 s^5 + b_1 s = 0 \rightarrow \mathbf{s=0}, \quad s_k = \left(\frac{b_1}{b_5}\right)^{1/4} e^{i(2k-1)\pi/4}$$

**We can write a general L-C impedance or admittance as**

$$Z(s) = \frac{K(s^2 + \omega_1^2)(s^2 + \omega_3^2)\dots(s^2 + \omega_i^2)\dots}{s(s^2 + \omega_2^2)(s^2 + \omega_4^2)\dots(s^2 + \omega_j^2)\dots}$$

$$Z(s) = \frac{K_0}{s} + \frac{2K_2s}{s^2 + \omega_2^2} + \frac{2K_4s}{s^2 + \omega_4^2} + \dots + K_\infty s$$

Since these poles are all on the  $j\omega$  axis, the residues must be real and positive in order for  $Z(s)$  to be positive real .

**$S=j\omega \rightarrow Z(j\omega)=jX(\omega)$  (no real part)**

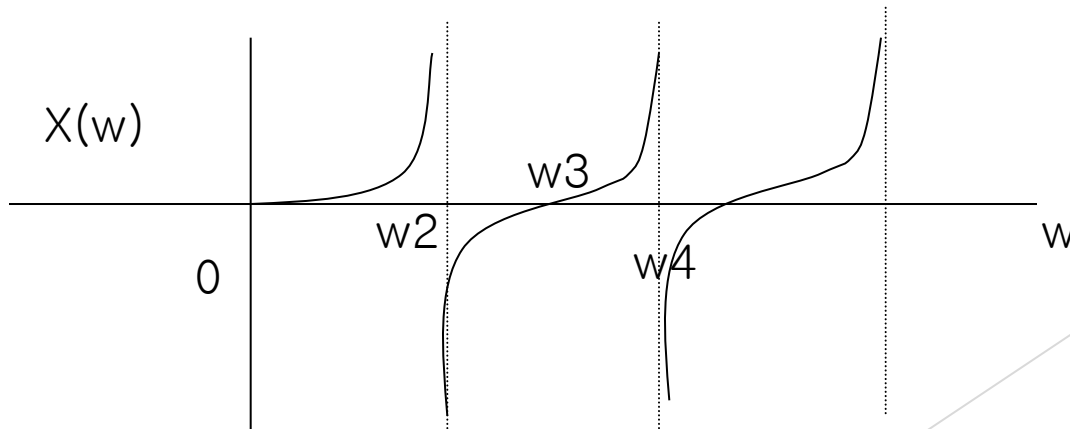


$$\frac{dX(\omega)}{d\omega} = \frac{K_0}{\omega^2} + K_\infty + \frac{K_2(\omega^2 + \omega_2^2)}{(\omega_2^2 + \omega^2)} + \dots$$

Since all the residues  $K_i$  are positive, it is easy to see that for an L-C function

$$\frac{dX(\omega)}{d\omega} \geq 0$$

$$\text{Ex) } Z(s) = \frac{Ks(s^2 + \omega_3^2)}{(s^2 + \omega_2^2)(s^2 + \omega_4^2)} \quad jX(\omega) = +j \frac{K\omega(-\omega^2 + \omega_3^2)}{(-\omega^2 + \omega_2^2)(-\omega^2 + \omega_4^2)}$$



# Properties 4 and 5. L-C immittance function

- ▶ The highest powers of the numerator and denominator must differ by unity; the lowest powers also differ by unity.
- ▶ There must be either a zero or a pole at the origin and infinity.

# Summary of properties

1.  $Z_{LC}(s)$  or  $Y_{LC}(s)$  is the ratio of odd to even or even to odd polynomials.
2. The poles and zeros are simple and lie on the  $j\omega$  axis
3. The poles and zeros interlace on the  $j\omega$  axis.
4. The highest powers of the numerator and denominator must differ by unity; the lowest powers also differ by unity.
5. There must be either a zero or a pole at the origin and infinity.

# Examples

$$Z(s) = \frac{Ks(s^2 + 4)}{(s^2 + 1)(s^2 + 3)}$$

$$Z(s) = \frac{s^5 + 4s^3 + 5s}{3s^4 + 6s^2}$$

$$Z(s) = \frac{K(s^2 + 1)(s^2 + 9)}{(s^2 + 2)(s^2 + 10)}$$

$$Z(s) = \frac{2(s^2 + 1)(s^2 + 9)}{s(s^2 + 4)}$$

# Synthesis of L-C Driving point immittances

- ▶ L-C immittance is a positive real function with poles and zeros on the  $j\omega$  axis only.

$$Z(s) = \frac{K_0}{s} + \frac{2K_2s}{s^2 + \omega_2^2} + \frac{2K_4s}{s^2 + \omega_4^2} + \dots + K_\infty s$$

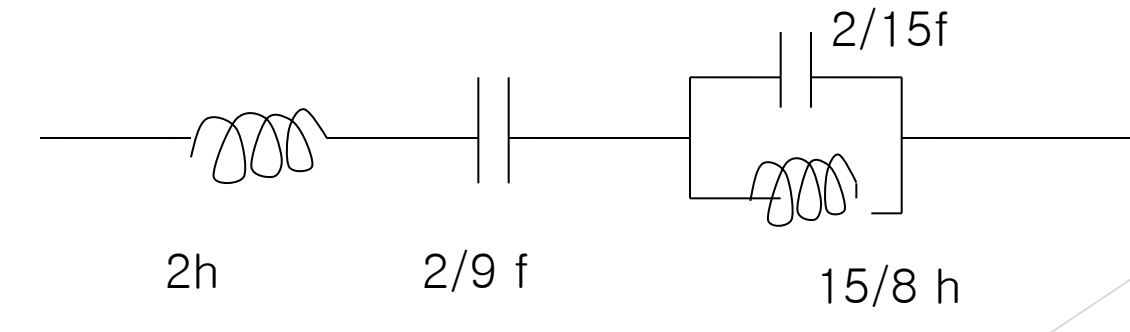
- ▶ The synthesis is accomplished directly from the partial fraction .
- ▶  $F(s)$  is impedance  $\rightarrow$  then the term  $K_0/s$  capacitor of  $1/K$  farads  
the  $K(\infty)s$  is an inductance of  $K(\infty)$  henrys.

# For $Z(s)$ partial fraction

$2K_i s / (s^2 + \omega_i^2)$  Is a parallel tank capacitance and inductance.

$$1/2K_i, \quad 2K_i / \omega_i^2$$

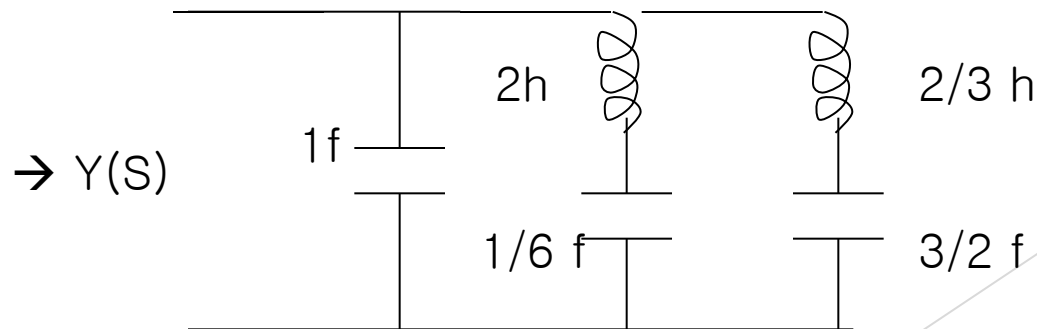
$$Z(s) = \frac{2(s^2 + 1)(s^2 + 9)}{s(s^2 + 4)} = 2s + \frac{9}{s} + \frac{15s}{s^2 + 4}$$



# For $Y(s)$ partial fraction

► In admittance

$$Y(s) = \frac{s(s^2 + 2)(s^2 + 4)}{(s^2 + 1)(s^2 + 3)} = s + \frac{\frac{1}{2}s}{s^2 + 3} + \frac{\frac{3}{2}s}{s^2 + 1}$$



# Another methodology

- ▶ Using property 4 “The highest powers of numerator and denominator must differ by unity; the lowest powers also differ by unity.”

- ▶ Therefore, there is always a zero or a pole at  $s = \infty$ .

- ▶ suppose  $Z(s)$  numerator:  $2n$ , denominator:  $2n-1$

- ▶ this network has pole at  $\infty$ .  $\rightarrow$  we can remove this pole by removing an impedance  $L_1 s$

$$Z_2(s) = Z(s) - L_1 s$$

- ▶ Degree of denominator :  $2n-1$  numerator:  $2n-2$

- ▶  $Z_2(s)$  has zero at  $s = \infty$ .

- ▶  $Y_2(s) = 1/Z_2(s)$ ,  $\rightarrow Y_3(s) = Y_2(s) - C_2 s$



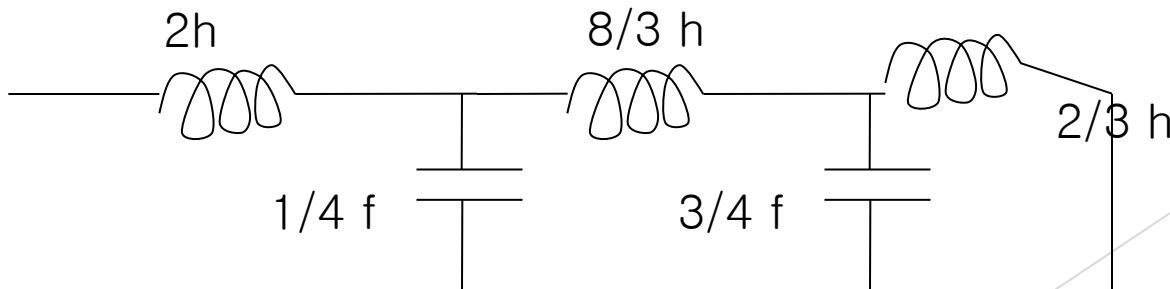
- ▶ This infinite term removing process continue until the remainder is zero.
- ▶ Each time we remove the pole, we remove an inductor or capacitor depending upon whether the function is an impedance or an admittance.
- ▶ → Final synthesized is a ladder whose series arms are inductors and shunt arms are capacitors.

$$Z(s) = \frac{2s^5 + 12s^3 + 16s}{s^4 + 4s^2 + 3}$$

$$Z_2(s) = \frac{2s^5 + 12s^3 + 16s}{s^4 + 4s^2 + 3} - 2s = \frac{4s^3 + 10s}{s^4 + 4s^2 + 3}$$

$$Y_3(s) = \frac{s^4 + 4s^2 + 3}{4s^3 + 10s} - \frac{1}{4}s = \frac{\frac{3}{2}s^2 + 3}{4s^3 + 10s}$$

$$Z_4(s) = Z_3 - \frac{8}{3}s = \frac{2s}{\frac{3}{2}s^2 + 3} \quad 2h$$



- ▶ This circuit (Ladder) called as Cauer because Cauer discovered the continues fraction method.
- ▶ Without going into the proof of the statement m in can be said that both the Foster and Cauer form gice the minimum number of elements for a specified L-C network.

# Example of Cauer Method

$$Z(s) = \frac{(s^2 + 1)(s^2 + 3)}{s(s^2 + 2)}$$

